

**SCS SLM TECHNIQUE FOR PAPR REDUCTION OF OFDM SIGNALS WITHOUT SIDE INFORMATION**

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**ABSTRACT**

Orthogonal frequency division multiplexing (OFDM) has been appeared and utilized as one of the center technology for the communication structures. In particular, OFDM has been followed for various wireless conversation systems which include wireless nearby location area networks (WLANs), Wireless metropolitan area network systems (WMANs), Digital Audio and also Video Broadcasting (DAB) and (DVB). OFDM is an engaging methodology for achieving high realities charge inside the Wi-Fi report frameworks and its miles vigorous to the recurrence particular blurring channel. Be that as it may, an OFDM sign can have peak-to-average power ratio (PAPR) at the transmitter, which reasons the signal mutilation, for example, the in-band and the out-of-band radiation because of the nonlinearity of highpower amplifier(HPA), and prompts the corruption of bit error rate (BER).As a consequence, the PAPR discount is solitary of the essential studies benefit for the OFDM systems. In this paper, we endorse a brand new PAPR reduction scheme of orthogonal frequency department multiplexing (OFDM) alerts and those are the brand new PTS schemes without aspect fact of side information (SI).

**Keywords:**

Orthogonal Frequency Division Multiplexing (OFDM), Partial Transmit Sequence (PTS), Peak-to-Average Power Ratio (PAPR), Phase offset, Side information (SI).

**INTRODUCTION**

OFDM has been followed as a popular technique in diverse Wi-Fi communiqué structures because it may reap high-price facts transmission and ensure high reliability over the multipath fading surroundings [1]. But, excessive top-to-common electricity ratio (PAPR) is an essential downside of OFDM due to the fact OFDM indicators are generated by summing a huge quantity of independently modulated subcarriers. Excessive PAPR reasons enormous inter-modulation and out-of-band radiation while OFDM indicators pass through nonlinear devices inclusive of the highpower amplifier (HPA).

To clear up the PAPR problem of OFDM indicators, many PAPR discount schemes were proposed consisting of clipping [2], [3], coding techniques [4], [5], selected mapping (SLM) [6]–[8], partial transmit sequence (PTS) [9]–[11], tone injection (TI) [12], tone reservation (TR) [12], [13] and active constellation extension (ACE) [14]. A low multifaceted nature PAPR discount for OFDM structures the usage of modified broadly linear SLM scheme is suggested in [9]. But this device does not obtain a considerable discount in complexity and the PAPR diminution performance is poorer compared to that of the traditional scheme. Novel selected mapping schemes with decreased complexity are evolved in [10]. Despite the fact that the computational complexity is substantially reduced, PAPR reduction performance is inferior to that of conventional SLM scheme. SLM scheme for a reduction in PAPR without the want of aspect facts is proposed in [12]. However, SLM method requires extra IFFT operations which increases the implementation complexity. Discount of PAPR in OFDM structures the use of Tone Reservation is proposed in [13–15] even as PAPR reduction the usage of clipping method is mentioned in [16–18].

PAPR discount strategies the uses of coding techniques are advised in [19–21]. Mahmudul Hasan proposed a way for PAPR discount in [19] the usage of linear predictive coding (LPC) which makes use of signal whitening belongings of LPC as a pre-processing step in OFDM systems. PAPR discount using Zad off-chu matrix rework (ZCMF) pre-coding based OFDM system is provided in [20] to permit the Radiofrequency amplifier to perform close to its saturation stage. PTS algorithm using Reed-Muller (RM) codes for blunders correction and PAPR discount is proposed in [21] which shows that RM codes in cyclic ordering gain higher performance than RM codes in natural ordering.

PAPR discount using various PTS techniques are mentioned in [2–7]. A decrease in PAPR of OFDM indicators employing partial transmit sequences is proposed in [2]. However, the conventional or traditional PTS algorithm calls for an exhaustive seek of the allowed segment element which increases the complexity. PAPR reduction using Interleaved PTS scheme which uses conjugate assets of DFT is usually recommended in [3].

On this paper, PTS schemes without SI are proposed, which carry out low-complexity detection of the selected rotating vector at the receiver. Much like the approach in [20], the proposed PTS schemes embed the SI identifying rotating vectors into the opportunity sign sequences by using giving identifiable segment offset to the factors of every pivoting vector.

In particular, the wide variety of phase offsets has to be nicely chosen through considering the number of subsequences of PTS scheme, and the phase rotating factors are section shifted by means of those offsets, that's unique to the scheme in [20] for an SLM case. To take out SI from the received symbol and recover the information collection, an ML detector is utilized. This ML indicator abuses the Euclidean separation between the given signal constellation and the signal constellation pivoted through the section offset. Via doing pairwise blunders chance (PEP) evaluation, it is investigated the way to choose proper segment counterbalances for embedding SI. Likewise, the overall performance degradation due to SI discovery disappointment is examined primarily based on PEP. thru numerical evaluation, it's miles proven that the BER performance of the proposed PTS schemes isn't degraded in comparison with the traditional PTS with best SI.

The rest of this paper is organized as follows. In Section 2, described the related works of PTS. Proposed work is reviewed and presented in Section 3. The proposed modified hybrid algorithm combining the additional hybrid with switching hybrid schemes is developed and analyzed in Section 4. Finally, some comparative simulation results are presented in Section 5, and the ultimate remarks are drawn in Section 6.

### RELATED WORKS

Wavelet Packet-based Orthogonal Frequency department Multiplexing (WP OFDM) systems have fascinating technique. However, the transmitted sign is be afflicted by excessive top-to-common power ratio (PAPR) because of the boom the supply of narrowband alerts that gather within the time domain [1]. An excessive PAPR occasion reasons non-linear distortion of indicators and reduces the power performance of an excessive-power amplifier (HPA).

Using an HPA is high in value and boom usual system complexity; as a result, PAPR discount is a good alternative. Specific strategies have been proposed to lessen PAPR, consisting of PTS with embedded aspect statistics [8], selected mapping [9], modified PTS [10], repeating clipping and filtering [11], greater PTS, particle swarm optimization (PSO), conventional PTS.

PAPR reduction techniques based totally on WP-OFDM systems, the convolutional-repeating clipping filtering PTS with embedded side statistics information is a good method to reduce PAPR without demanding signal orthogonally. The CRCF is a mixture of convolution code and repeating clipping and filtering. The convolution code is worn to encode the records and repeating clipping filtering are used to cut the high energy value of the information and filtering are used to reject the out of hand signals [2].

The repetition based technique incorporates coding, selective mapping (SLM), tone infusion (TI), fractional transmit groupings (PTS) and tone reservation (TR) and [7-13], and numerous others. For the excess based strategy, the undesired results going ahead to the contortion basically based procedures can be lightened in the meantime as the punishment is the decreased transmission cost or increased normal vitality because of the presentation of repetition.

A block coding strategy [14] is to transmit simplest the code phrases with low PAPR. Such coding strategies offer precise PAPR reduction performance and coding gain. Widespread boost of the coding technique for PAPR manage the usage of generalized Reed-Muller codes is summarized in [15]. The crucial hassle for the coding technique is that for the OFDM device with a massive number of subcarriers, both it encounters layout difficulties or the consequent coding rate turns into prohibitively low.

Contrary to the schemes in [15] and [16], the proposed PTS schemes do not adjust the pilot tones, that is, both spectral performance and correct channel estimation is achieved through the proposed PTS schemes. Additionally, the proposed schemes do not now extend constellation points as completed in [17]. As a substitute, the anticipated schemes rotate the sign constellation by methods for the utilization of a small number of segment offsets. Hence, the proposed schemes display low computational complexity to look into a smaller search area compared with the schemes in [18] and [19].

### PROPOSED SYSTEM

PAPR deduction in OFDM system using new PTS scheme without the side information. In the commencement, the conventional hybrid (CH) method combining the PTS schemes with SLM technique is investigated. OFDM is a unique kind of multi-carrier transmission method that divides the verbal exchange channel into many equally spaced frequency bands. The concept of Partial Transmit series (PTS) method is carried out to the OFDM signs to lessen excessive height signals. Coding and simulation are done for PTS and their outcomes on decreasing the PAPR are discovered. Partial Transmit Sequence (PTS) set of rules is a procedure for enhancing the facts of a multicarrier signal [7]. The fundamental thoughts of partial transmit sequences algorithm is to divide the unique OFDM series into several sub-sequences and for each sub-sequences improved by way of exclusive weights until a greatest cost is chosen (fig 1) from the left facet of diagram, the statistics information in frequency area x is isolated into v non covering sub-pieces of size n [11], that every n/v nonzero factors and set the rest part to zero.. These sub-pieces are accepted to have the identical length and no hole between each different.

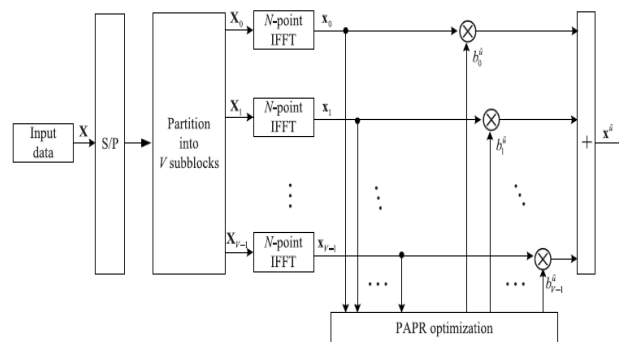


Fig. 1. Block diagram of the conventional PTS Scheme.

### PAPR PROBLEM AND SLM SCHEME

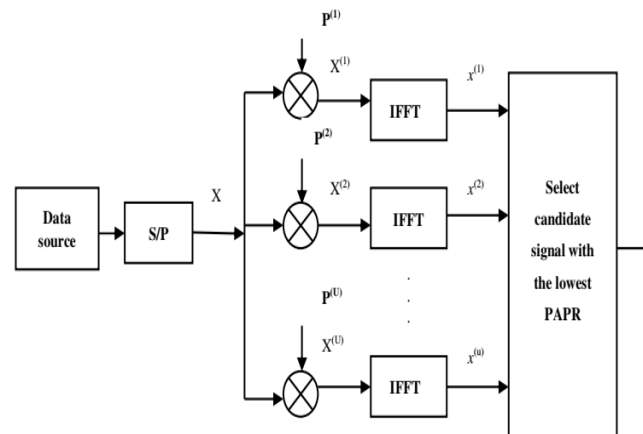
In the discrete time space, an OFDM flag  $x_n$ , of N subcarriers can be communicated,

$$x_n = \frac{1}{\sqrt{N}} \sum X_k e^{\frac{j2\pi kn}{N}}, 0 \leq n \leq N - 1 \tag{1}$$

Where  $X_k$ ,  $k = 0, 1, 2, 3, \dots, N-1$  are input signs modulated by BPSK, QPSK or QAM and n is the discrete time index. The PAPR of an OFDM signal is characterized as the proportion of the most extreme to the average power of the signal, as follows,

$$\text{PAPR}(x) = 10 \log_{10} \frac{\max\{|x_n|^2\}}{E\{|x|^2\}} \tag{2}$$

Where  $E \{ \cdot \}$  indicates the expected value operation and  $x = [x_1, x_2, x_3, x_4, \dots, x_{N-1}]^T$ . As to the discrete-time signals, seeing that symbol-spaced sampling may additionally every now and then miss a number of the signal peaks, sign samples are received by means of oversampling by an issue of  $L$  to better inexact the proper PAPR. Oversampled time-domain samples are usually obtained via  $L$ -factor IFFT of the information block with  $(L-1)$   $N$  0-padding. It is proven in [20] that  $L = \text{four}$  is sufficient to seize the peaks. At the point when the OFDM motion with high PAPR goes through a non-linear device, (energy amplifier working within the saturation area), the sign will go through vast non-linear distortion [20].



*Fig. 2. Block diagram of the SCSSLM scheme*

### IMPLEMENTATION OF HYBRID SCHEME

The point of this area is two partial transmit grouping (PTS) plans lacking side information (SI) are proposed for decreasing PAPR proportion of OFDM signals. The proposed plans don't transmit SI recognizing a pivoting vector in light of the fact that identifiable stage counterbalance is connected to the components of each turning vector. To remove SI from the conventional signal and recover the data gathering, the Maximum probability (ML) discoverer is used. This ML locator misuses the Euclidean separation between the given flag group of stars and the flag heavenly body turned by the stage balance.

By liability pairwise error likelihood (PEP) examination, it is researched how to pick great stage counterbalances for installing SI.

### OFDM

OFDM has been regarded and used as one of the core technologies for the communication systems. Especially, OFDM has been embraced for different remote correspondence frameworks, for example, Remote Wireless LocalArea Networks (WLANs), Digital Audio and Video Broadcasting (DAB) and (DVB), remote metropolitan Area Network systems (WMANs). OFDM is an appealing procedure for accomplishing high information rate in the remote correspondence frameworks and it is strong to the recurrence particular fading channel. In any case, an OFDM flag can have high Peak to-normal power proportion (PAPR) at the transmitter, which causes the flag contortion, for example, the in-band mutilation and the out-of-band radiation because of the nonlinearity of high power amplifiers (HPA), and instigates the debasement of bit error rate (BER). Henceforth, the PAPR diminish is a champion among the most essential research interests for the OFDM structures.

### PTS Scheme

The point of this area is two halfway transmit succession (PTS) plans without side data (SI) are proposed for diminishing PAPR proportion of orthogonal Frequency division multiplexing signals. The proposed plans don't transmit SI recognizing a pivoting vector in light of the fact that identifiable stage counterbalance is connected to the components of each turning vector. To separate SI from the got flag and recuperate the information arrangement, the Maximum probability (ML) indicator is utilized. This ML finder misuses the Euclidean separation between the given flag star grouping and the flag heavenly body turned by the stage balance.

By doing pairwise error probability (PEP) examination, it is explored how to pick great stage balances for installing SI.

### Embedding Side Information into Rotating Vectors

Vectors To dispose of the transmission of SI, we propose a strategy to implant the SI into the turning vectors through stage moving them by fitting counterbalances. Initial, a V-tuple stage counterbalance vector is characterized as

$$S^u = [S_0^u S_1^u \dots S_{V-1}^u] \dots\dots\dots(3)$$

$$\bar{b}^u = [\bar{b}_0^u \bar{b}_1^u \dots \bar{b}_{V-1}^u] = [b_0^u e^{j\theta_{s_0^u}} b_1^u e^{j\theta_{s_1^u}} \dots b_{V-1}^u e^{j\theta_{s_{V-1}^u}}]$$

$$x^u = \sum_{v=0}^{V-1} \bar{b}_v^u x_v \dots\dots\dots(4)$$

### ML Detection for P-PTS I

From the signal that is received and generated by the modified pivoting vector  $\bar{b}^u$ , the receiver should find the index  $\tilde{u}$  without using SI and recover the input symbol sequence X. The received  $R_k$  at the k-th subcarrier, which belongs to the v-th symbol subsequence  $X_v$ , in the frequency domain is expressed as

$$R_k = H_k \bar{b}_v^u X_k + N_k \dots\dots\dots(5)$$

where  $H_k$  is the frequency response and  $N_k$  is an AWGN sample at the k-th subcarrier with the variance per dimension  $N_0/2$ . It is assumed that channel is quasi-static Rayleigh fading channel and  $H_k$ 's are statistically independent and perfectly known to the receiver, that is, the perfect channel state information (CSI) is assumed.

### Design Criteria Of V-tuple Phase Offset Vectors $S_u$ And Phase Offsets

In this subsection, design criteria of V-tuple phase offset vectors  $S_u$  and phase offsets are derived, which minimize the PEP of ML detector. Note that PEP analysis is widely used for analyzing the error correction performance of communication systems with coding schemes such as trellis codes and space-time codes. For the convenience of PEP analysis, we assume that the adjacent sub-block partitioning is used and the similar analysis can be applied to other sub-block partitioning methods. Consider that every component of the signal constellation is shrunk by a scale factor  $\sqrt{E_s}$  with the goal that the normal vitality signal constellation is 1, where  $E_s$  is the normal vitality of the transmitted symbols.

### Design of V-Tuple Phase Offset Vectors and Phase Offsets

If the number of nonzero phase offsets Z is large, a set of V-tuple phase offset vectors having large minimum Hamming distance following the first criterion can be easily constructed. However, as Z increases, the average Euclidean separation between the original and rotated signal constellations is considerably reduced, which violates the second criterion, and the detection complexity of the proposed ML detector also increases. Therefore, it is difficult to find the optimal V-tuple phase offset vectors and the optimal phase offsets to simultaneously satisfy two design criteria. Through the extensive numerical analysis of the viewpoints of BER performance and the implementation.

### Performance Degradation by SI Detection Failure

The disappointment likelihood of SI mainly depends on SNR. In this subsection, we describe the relationship between them. Similar to the derivation of PEP for the proposed ML since  $N/V \gg 1$  in the OFDM system,

$$(E[|\hat{X}_k - \hat{X}_k|^2])^{\frac{1}{\sqrt{V} \frac{N}{V}}} \approx 1 \dots\dots\dots(2)$$

Therefore, the approximation is rewritten as  $\approx \frac{4}{E_s/N_0} - E[|A_k|^2]$

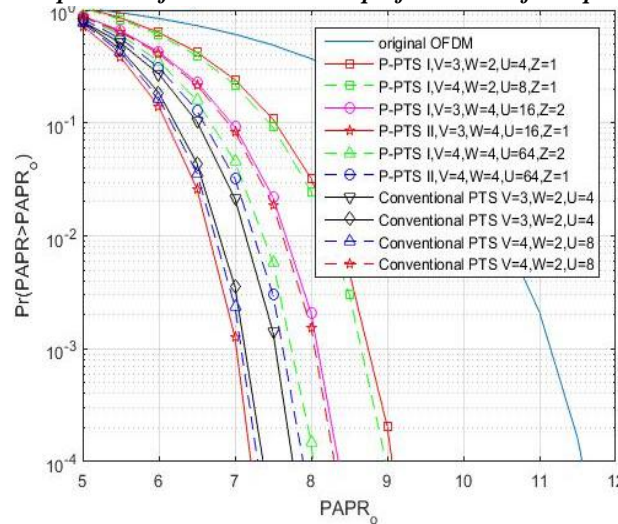
where depends on the number of phase offsets and used constellations. If it is larger than 0, we can state that the disappointment likelihood of SI mainly degrades the bit error performance. On the other hand, it is smaller than 0, we can state that the disappointment likelihood of SI is not the main factor of bit error compared to the channel noise. That is, in low SNR region the failure probability of SI mainly degrades the bit error performance of the OFDM system.

### SIMULATION RESULT

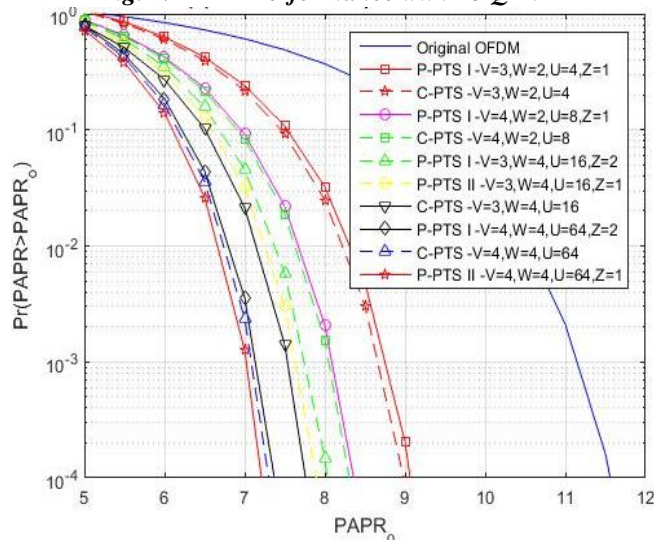
The proposed Wavelet Packet-based CRCF-PTS with embedded side information data is simulated to analyze and examine its overall performance with wavelet packet based PTS with ESID by using the usage of MATLAB device. It is investigated the PAPR reduction may be executed thru the proposed scheme. A major realistic drawback discovered in MCM transmission approach together with the traditional OFDM is the yield of random envelops with excessive height.

When the high top sign operates within the non-linear area of the electricity amplifier on the transmitter, the system experiences non-linear distortion. The WP based totally PTS associated works brought to successfully lessen the excessive PAPR with the expense of applicable BER overall performance degradation. But, their strategies ignore the SI data. In this work, the SI records are delivered to the body and also the usage of convolutional repeating clipping filtering approach on the transmitter reduces the PAPR decreased and the BER overall performance is advanced. Therefore, this serves because of the motivation for acquiring the consequences of the proposed scheme.

**Figure 3. Comparison of PAPR reduction performance of two proposed PTS**



**Fig. 4. PAPR Performance with 16 QAM**



**Fig. 5. Comparison of BER of two proposed PTS schemes and the conventional PTS scheme with PSI**

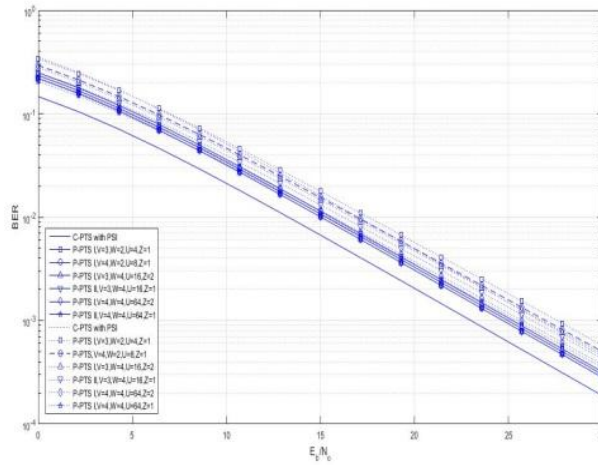


Fig. 6. The PAPR reduction performance of switching hybrid scheme for OFDM systems.

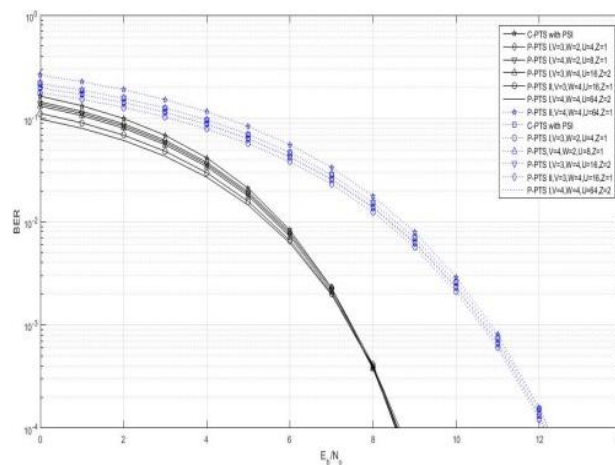


Fig. 7. Comparison of BER of two proposed PTS schemes and the conventional PTS scheme with PSI

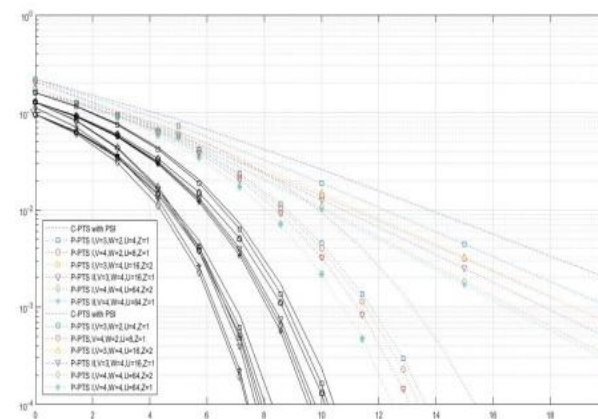
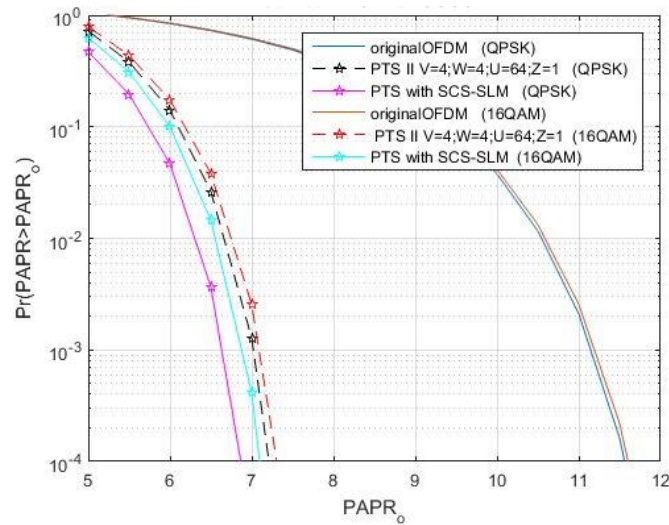


Figure 8. PAPR performance for traditional PTS and proposed PTS schemes



**Table 1. PAPR Reduction Performance - QPSK**

| Scheme   | BER       | SNR in dB |
|----------|-----------|-----------|
| P-PTS I  | $10^{-3}$ | 9         |
| P-PTS II | $10^{-3}$ | 8.2       |
| C-PTS    | $10^{-3}$ | 7.2       |

**Table 2. PAPR Reduction Performance - QAM**

| Scheme   | BER       | SNR Ratio |
|----------|-----------|-----------|
| P-PTS I  | $10^{-3}$ | 8         |
| P-PTS II | $10^{-3}$ | 7.9       |
| C-PTS    | $10^{-3}$ | 7.2       |

**Table 3. PAPR Reduction Performance – SCS SLM**

| Scheme                  | BER       | SNR Ratio |
|-------------------------|-----------|-----------|
| P-PTS with SCS SLM QPSK | $10^{-3}$ | 6.9       |
| P-PTS 16 QAM            | $10^{-3}$ | 7.1       |

**Table 4. PAPR Reduction Performance – SCS SLM**

| Scheme                  | SNR Ratio | BER                |
|-------------------------|-----------|--------------------|
| P-PTS with SCS SLM QPSK | 6         | $10^{-1.4}$ =0.398 |
| P-PTS 16 QAM            | 6         | $10^{-1}$ =0.1     |



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